

Lecture 13 - Observers

K. J. Åström

1. Introduction
2. The Observer Problem
3. Combining with State Feedback
4. Integral action
5. Response to Reference Signals
6. Summary

Theme: Estimating the state and using the estimate for feedback.

Introduction

- The white box view of dynamical systems
- *State* a number of variables that summarizes the past that is useful for prediction
- The pole placement can be solved if all states are measured
- What to do if only outputs are known?
- How to determine the state from the output?
- Combine it with state feedback
- A new view on integral action

Computing the State from the Output

System

$$\begin{aligned}\frac{dx}{dt} &= Ax \\ y &= Cx\end{aligned}$$

$$\begin{pmatrix} C \\ CA \\ \vdots \\ CA^{n-1} \end{pmatrix} x = W_o x = \begin{pmatrix} y \\ \frac{dy}{dt} \\ \vdots \\ \frac{d^{n-1}y}{dt^{n-1}} \end{pmatrix}$$

Differentiate the output

$$\begin{aligned}y &= Cx \\ \frac{dy}{dt} &= CAx \\ &\vdots \\ \frac{d^{n-1}y}{dt^{n-1}} &= CA^{n-1}x\end{aligned}$$

W_o observability matrix

$$x = W_o^{-1} \begin{pmatrix} y \\ \frac{dy}{dt} \\ \vdots \\ \frac{d^{n-1}y}{dt^{n-1}} \end{pmatrix}$$

An Other Attempt

The problem can be solved if the observability matrix has full rank. Can we avoid differentiations?

$$\begin{aligned}\frac{dx}{dt} &= Ax + Bu \\ y &= Cx\end{aligned}$$

Simulate a model of the system

$$\frac{d\hat{x}}{dt} = A\hat{x} + Bu$$

Introduce error $\tilde{x} = x - \hat{x}$

$$\frac{d\tilde{x}}{dt} = A(x - \hat{x}) = A\tilde{x}$$

Works if A stable, but it makes no use of u .

Two Ways to Calculate the State

By differentiating the output

$$x = W_o^{-1} \left(y \quad \frac{dy}{dt} \quad \dots \quad \frac{d^{n-1}y}{dt^{n-1}} \right)^T$$

This method has the drawback that it only uses the output and that it is differentiated many times which is highly sensitive to noise.

By integrating the input

$$\frac{d\hat{x}}{dt} = A\hat{x} + Bu$$

This method has the drawback that it only uses the input and that it only works if the matrix A is stable.

Can we combine the methods?

An Alternative Solution

$$\begin{aligned} \frac{dx}{dt} &= Ax + Bu \\ y &= Cx \end{aligned}$$

Drive the model by both u and y

$$\frac{d\hat{x}}{dt} = A\hat{x} + Bu + K(y - C\hat{x})$$

Introduce error $\tilde{x} = x - \hat{x}$

$$\frac{d\tilde{x}}{dt} = A(x - \hat{x}) - K(y - C\hat{x}) = (A - KC)\tilde{x}$$

Notice that use of output y gives extra freedom.

Determine matrix K so that the matrix $A - KC$ has all its eigenvalues in the left half plane, then \tilde{x} will go to zero!

Example - The Car

$$\begin{aligned} \frac{dx}{dt} &= Ax + Bu = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} x + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u \\ y &= Cx = \begin{pmatrix} 1 & 0 \end{pmatrix} x \end{aligned}$$

Observability matrix

$$W_o = \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix}$$

has full rank. The system is observable and the observer is

$$\frac{d\hat{x}}{dt} = A\hat{x} + Bu = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} \hat{x} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u + \begin{pmatrix} k_1 \\ k_2 \end{pmatrix} (y - \hat{x}_1)$$

$$\text{We have } A - KC = \begin{pmatrix} -k_1 & 1 \\ -k_2 & 0 \end{pmatrix}$$

Example - The Car ...

The matrix

$$A - KC = \begin{pmatrix} -k_1 & 1 \\ -k_2 & 0 \end{pmatrix}$$

has the characteristic polynomial

$$\det(sI - A + KC) = \det \begin{pmatrix} s + k_1 & -1 \\ k_2 & s \end{pmatrix} = s^2 + k_1s + k_2$$

Choosing

$$\begin{aligned} k_1 &= 2\zeta\omega_o \\ k_2 &= \omega_o^2 \end{aligned}$$

gives the characteristic polynomial $s^2 + 2\zeta\omega_o s + \omega_o^2$.

Interpretation of the Observer

The observer is a dynamical system with two inputs u and y and two outputs x_1 and x_2 .

$$\frac{d\hat{x}}{dt} = \begin{pmatrix} 0 & 1 \\ 0 & 0 \end{pmatrix} \hat{x} + \begin{pmatrix} 0 \\ 1 \end{pmatrix} u + \begin{pmatrix} k_1 \\ k_2 \end{pmatrix} (y - \hat{x}_1)$$

Input-output relations

$$\begin{aligned} \hat{X}_1(s) &= \frac{1}{s^2 + 2\zeta\omega_o s + \omega_o^2} U(s) + \frac{2\zeta\omega_o s + \omega_o^2}{s^2 + 2\zeta\omega_o s + \omega_o^2} Y(s) \\ \hat{X}_2(s) &= \frac{s + 2\zeta\omega_o}{s^2 + 2\zeta\omega_o s + \omega_o^2} U(s) + \frac{\omega_o^2 s}{s^2 + 2\zeta\omega_o s + \omega_o^2} Y(s) \end{aligned}$$

Physical interpretation. Compare with $\hat{x}_2 = \frac{dy}{dt}$

Comparison with State Feedback

State feedback design: Find matrix L so that the matrix $A - B\mathbf{L}$ has prescribed eigenvalues.

Observer design: Find Matrix K so that the matrix $A - \mathbf{K}C$ has prescribed eigenvalues.

A matrix and its transpose have the same eigenvalues. We have

$$(A - \mathbf{K}C)^T = A^T - C^T \mathbf{K}^T$$

State feedback and observer design are the same problem. The same software can be used. Use state feedback program to obtain the observer gain by feeding it by A^T and C^T instead of A and B . Transposing L give the observer gain.

Observable Canonical Form

$$\begin{aligned} \frac{dz}{dt} &= \begin{pmatrix} -a_1 & 1 & 0 & \dots & 0 \\ -a_2 & 0 & 1 & & 0 \\ \vdots & & & & \\ -a_{n-1} & 0 & 0 & & 1 \\ -a_n & 0 & 0 & & 0 \end{pmatrix} z + \begin{pmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{pmatrix} u \\ y &= \begin{pmatrix} 1 & 0 & 0 & \dots & 0 \end{pmatrix} z \end{aligned}$$

$$\tilde{W}_o = \begin{pmatrix} 1 & 0 & 0 & \dots & 0 \\ -a_1 & 1 & 0 & \dots & 0 \\ a_1^2 - a_2 & -a_1 & 1 & \dots & 0 \\ \vdots & & & & \\ . & . & . & \dots & 1 \end{pmatrix}, \quad \tilde{W}_o^{-1} = \begin{pmatrix} 1 & 0 & \dots & 0 \\ -a_1 & 1 & \dots & 0 \\ -a_2 & -a_1 & \dots & 0 \\ \vdots & & & \\ -a_n & -a_{n-1} & \dots & 1 \end{pmatrix}$$

Observable Canonical Form

$$\begin{aligned} \frac{dz}{dt} &= \begin{pmatrix} -a_1 & 1 & 0 & \dots & 0 \\ -a_2 & 0 & 1 & & 0 \\ \vdots & & & & \\ -a_{n-1} & 0 & 0 & & 1 \\ -a_n & 0 & 0 & & 0 \end{pmatrix} z + \begin{pmatrix} b_1 \\ b_2 \\ \vdots \\ b_n \end{pmatrix} u \\ y &= \begin{pmatrix} 1 & 0 & 0 & \dots & 0 \end{pmatrix} z \end{aligned}$$

$$\frac{d\hat{z}}{dt} = \begin{pmatrix} -a_1 & 1 & 0 & \dots & 0 \\ -a_2 & 0 & 1 & & 0 \\ \vdots & & & & \\ -a_{n-1} & 0 & 0 & & 1 \\ -a_n & 0 & 0 & & 0 \end{pmatrix} \hat{z} + \begin{pmatrix} b_1 \\ b_2 \\ \vdots \\ b_{n-1} \\ b_n \end{pmatrix} u + \begin{pmatrix} k_1 \\ k_2 \\ \vdots \\ k_{n-1} \\ k_n \end{pmatrix} (y - \hat{z}_1)$$

Summary

If the system

$$\begin{aligned}\frac{dx}{dt} &= Ax \\ y &= Cx\end{aligned}$$

is observable its state can be determined by the observer

$$\frac{d\hat{x}}{dt} = A\hat{x} + Bu + K(y - C\hat{x})$$

- Combine measurements with mathematical model - indirect measurements -redundancy
- Sensor fusion in AI terminology
- Determination of filter gain K similar to state feedback design. Relation to the Kalman filter

Output Feedback

Consider a system which is observable and controllable

$$\begin{aligned}\frac{dx}{dt} &= Ax \\ y &= Cx\end{aligned}$$

Determine the state with an observer and use state feedback from the observed state. The controller is

$$\begin{aligned}u &= -L\hat{x} \\ \frac{d\hat{x}}{dt} &= A\hat{x} + Bu + K(y - C\hat{x})\end{aligned}$$

The controller is a dynamical system whose dynamics is represented by the observer

The Closed Loop System

$$\begin{aligned}\frac{dx}{dt} &= Ax + Bu \\ y &= Cx \\ u &= -L\hat{x} \\ \frac{d\hat{x}}{dt} &= A\hat{x} + Bu + K(y - C\hat{x})\end{aligned}$$

Introduce the state $\tilde{x} = x - \hat{x}$ instead of \hat{x} .

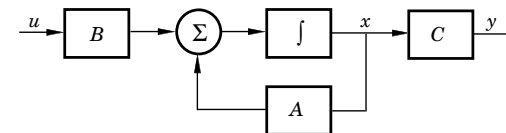
$$\begin{aligned}\frac{dx}{dt} &= Ax + Bu = Ax - BL\hat{x} = Ax - BL(x - \tilde{x}) = (A - BL)x + BL\tilde{x} \\ \frac{d\tilde{x}}{dt} &= (A - KC)\tilde{x}\end{aligned}$$

Characteristic equation

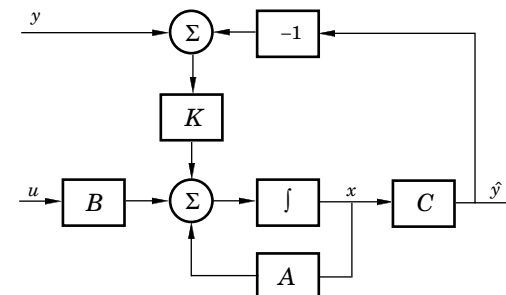
$$\det(sI - A + BL) \det(sI - A + KC) = 0$$

Block Diagram of Process and Observer

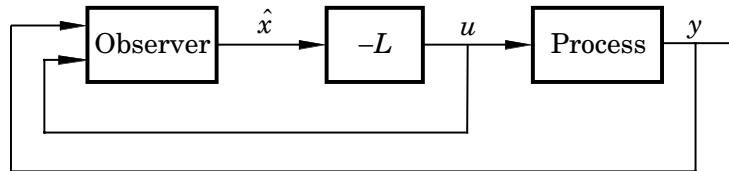
Process



Observer contains a copy of the process model



Block Diagram of Closed Loop system



A few things remains

- Integral action
- How to introduce reference signals

Observers and State Feedback - A Summary

Design of a controller for

$$\frac{dx}{dt} = Ax + Bu, \quad y = Cx$$

can be split into two problems

- Design of a state feedback and an observer
- The closed loop system has poles corresponding to the eigenvalues of the state feedback $A_c = A - BL$ and the observer $A_o = A - KC$ matrices
- The problems are similar. The observer is obtained by the transformation $A \rightarrow A^T$, $B \rightarrow C^T$ and $L \rightarrow K^T$. The same computer program can be used.
- Interesting interpretation of controller

Integral Action by Explicit Disturbance Modeling

There are many ways to introduce integral action. Here we will give a method based on explicit modeling of disturbances.

- What is the meaning of the model

$$\frac{dx}{dt} = Ax + Bu$$

- Why do we introduce integral action?
- Disturbances!
- How to describe disturbances

Modeling Disturbances

Classical disturbance models:

- Step $\frac{dv}{dt} = 0, \quad v(t) = a = \text{constant}$
- Ramp $\frac{d^2v}{dt^2} = 0, \quad v(t) = a + bt$
- Sinusoid $\frac{d^2v}{dt^2} + \omega^2 v = 0, \quad v(t) = a \sin(\omega t + b)$

Constant Disturbance at Process Input

$$\begin{aligned}\frac{dx}{dt} &= Ax + B(u + v) \\ \frac{dv}{dt} &= 0\end{aligned}$$

Feedback from estimated states

$$u = -L\hat{x} - \hat{v}$$

Observer

$$\begin{aligned}\frac{d}{dt} \begin{pmatrix} \hat{x} \\ \hat{v} \end{pmatrix} &= \begin{pmatrix} A & B \\ 0 & 0 \end{pmatrix} \begin{pmatrix} \hat{x} \\ \hat{v} \end{pmatrix} + \begin{pmatrix} B \\ 0 \end{pmatrix} u + \begin{pmatrix} K \\ K_v \end{pmatrix} \varepsilon \\ \varepsilon &= (y - C\hat{x})\end{aligned}$$

Example - The Car

Constant disturbance $x_3 = v$, physical interpretation.

$$\begin{aligned}\frac{dx}{dt} &= \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{pmatrix} x + \begin{pmatrix} 0 \\ 1 \\ 0 \end{pmatrix} u \\ y &= Cx = (1 \ 0 \ 0) x\end{aligned}$$

System is observable (check this!). The observer

$$\frac{d\hat{x}}{dt} = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{pmatrix} \hat{x} + \begin{pmatrix} 0 \\ 1 \\ 0 \end{pmatrix} u + \begin{pmatrix} k_1 \\ k_2 \\ k_3 \end{pmatrix} (y - \hat{x}_1)$$

Example - The Car ...

The matrix

$$A - KC = \begin{pmatrix} -k_1 & 1 & 0 \\ -k_2 & 0 & 1 \\ -k_3 & 0 & 0 \end{pmatrix}$$

has the characteristic polynomial

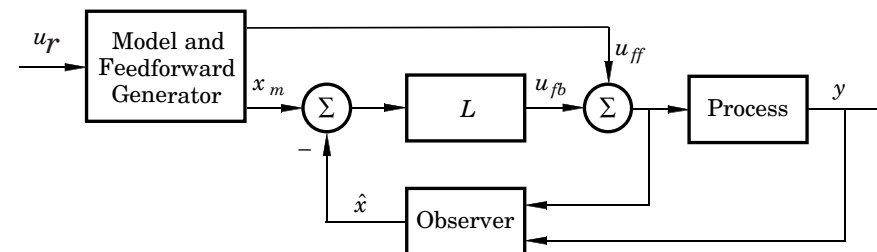
$$\det(sI - A + KC) = \begin{vmatrix} s + k_1 & -1 & 0 \\ k_2 & s & -1 \\ k_3 & 0 & s \end{vmatrix} = s^3 + k_1 s^2 + k_2 s + k_3$$

choose $k_1 = a + 2\zeta\omega_o$, $k_2 = 2a\zeta\omega_o + \omega_o^2$, and $k_3 = a\omega_o^2$ to give characteristic polynomial

$$(s + a)(s^2 + 2\zeta\omega_o s + \omega_o^2) = s^3 + (a + 2\zeta\omega_o)s^2 + (2a\zeta\omega_o + \omega_o^2)s + a\omega_o^2$$

How to Introduce Reference Signals

To introduce the reference values we use the standard configuration with two degrees of freedom.



- A nice structure
- Decoupling of estimation of disturbances K
- Reduction of disturbances L
- Response to reference signals

Example - The Car

$$\frac{dx}{dt} = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{pmatrix} x + \begin{pmatrix} 0 \\ 1 \\ 0 \end{pmatrix} u$$

$$y = Cx = (1 \ 0 \ 0)x$$

Desired response in Laplace transforms

$$Y_m = X_{m1} = M_y R = \frac{\omega_m^2}{s^2 + 2\zeta\omega_m s + \omega_m^2} R, \quad X_{m2} = sX_{m1}$$

Hence

$$U_m = P^{-1}M_y R = \frac{\omega_m^2 s^2}{s^2 + 2\zeta\omega_m s + \omega_m^2} R$$

$$= \omega_m^2 \left(1 - \frac{2\zeta\omega_m s + \omega_m^2}{s^2 + 2\zeta\omega_m s + \omega_m^2}\right) R = \omega_m^2 \left(R - \frac{2\zeta}{\omega_m} s Y_m - Y_m\right)$$

Example - The Car ...

The complete controller

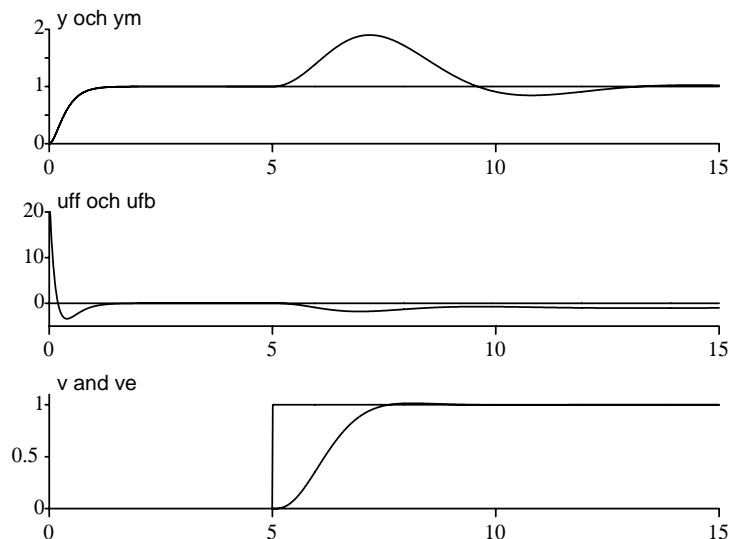
$$u = l_1(x_{m1} - \hat{x}_1) + l_2(x_{m2} - \hat{x}_2) - \hat{x}_3 + u_m$$

$$u_m = \omega_m^2 \left(r - \frac{2\zeta\omega_m}{\omega_m^2} x_{m2} - x_{m1}\right)$$

$$\frac{dx_m}{dt} = \begin{pmatrix} 0 & 1 \\ -\omega_m^2 & -2\zeta\omega_m \end{pmatrix} x_m + \begin{pmatrix} 0 \\ \omega_m^2 \end{pmatrix} r$$

$$\frac{d\hat{x}}{dt} = \begin{pmatrix} 0 & 1 & 0 \\ 0 & 0 & 1 \\ 0 & 0 & 0 \end{pmatrix} \hat{x} + \begin{pmatrix} 0 \\ 1 \\ 0 \end{pmatrix} u + \begin{pmatrix} k_1 \\ k_2 \\ k_3 \end{pmatrix} (y - \hat{x}_1)$$

A Simulation



Summary

- An elegant approach to control system design
- Interesting system structure with three components:
 - state feedback
 - observer
 - reference trajectory generator
- Strong similarity between state feedback and observers
- Sensor fusion and diagnosis $y - \hat{y}$
- A new interpretation of integral action: disturbance estimator. Can be adapted to other signals.